

Scientific Computing, Utrecht, March 24, 2014

# Fourier Transforms Wavelets Theory and Applications

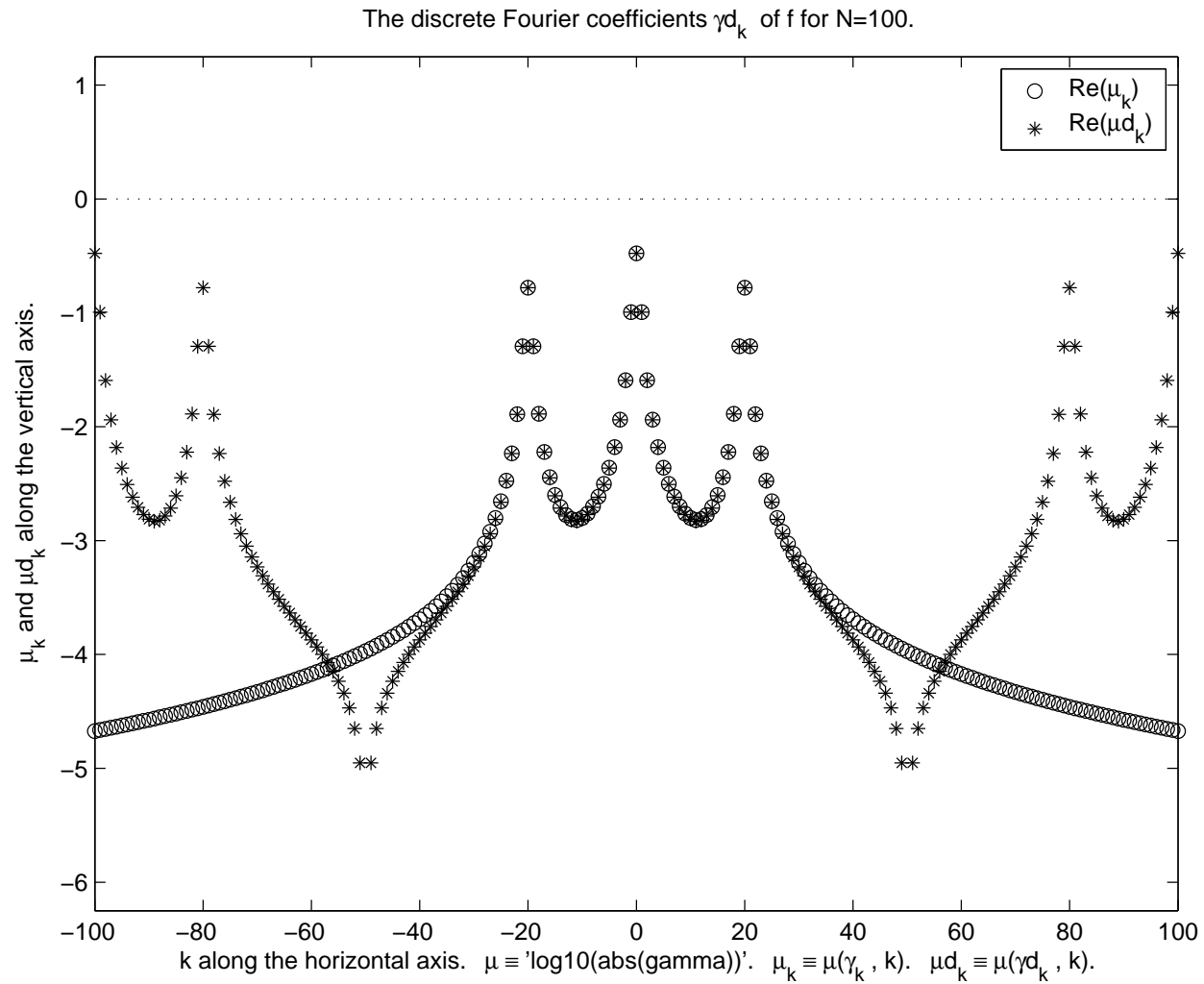
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# Computing Fourier Transforms



# Program

- Computing Fourier Coefficients
- Discrete Fourier Transform
- Discrete Cosine Transform
- Fast Fourier Transform
- Computing Fourier Integrals

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## Fourier coefficients

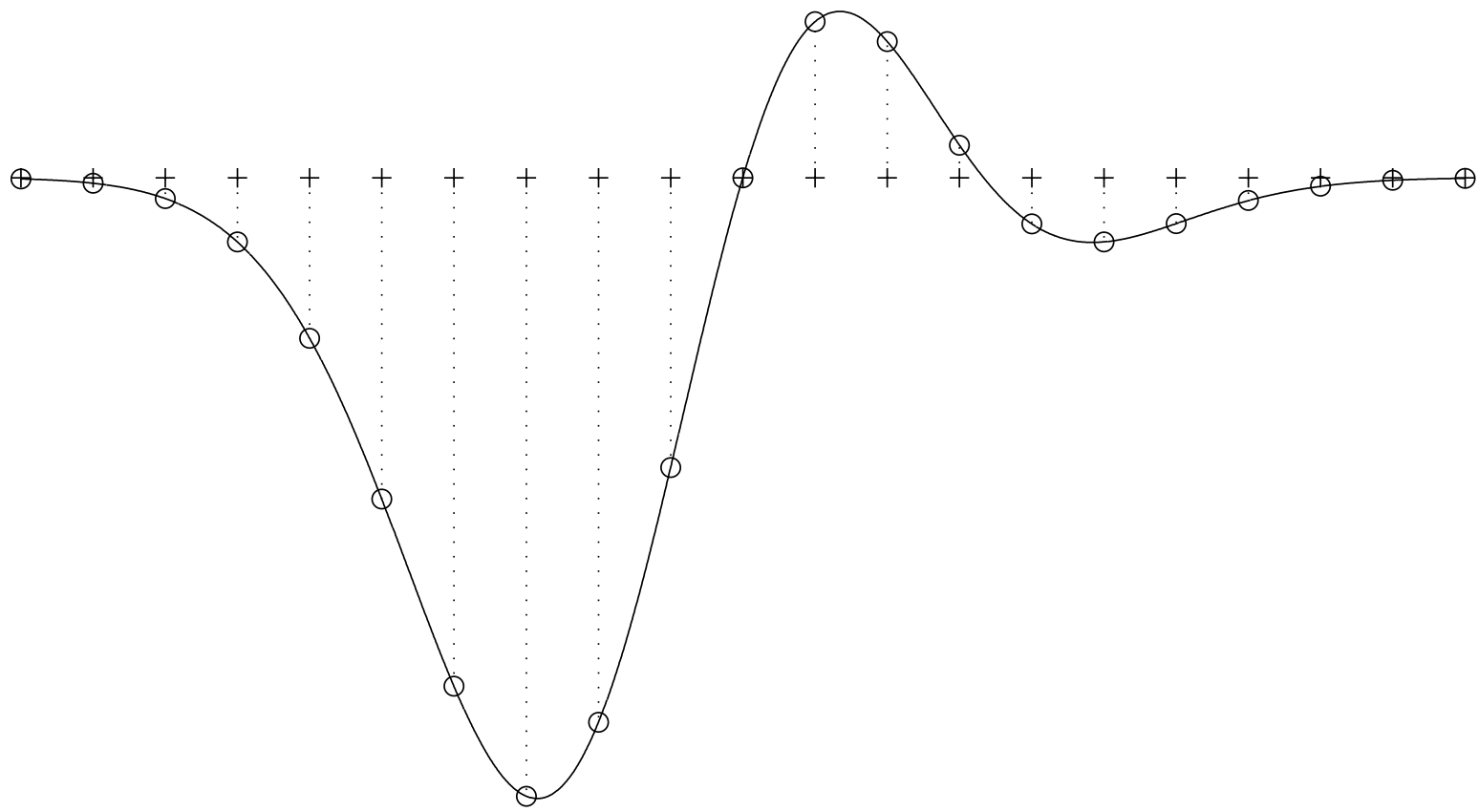
Let  $f : \mathbb{R} \rightarrow \mathbb{C}$  be  $T$ -periodic and sufficiently smooth.

$$\gamma_k(f) = \frac{1}{T} \int_0^T f(t) e^{-2\pi i t \frac{k}{T}} dt, \quad f(t) = \sum_{k \in \mathbb{Z}} \gamma_k(f) e^{2\pi i t \frac{k}{T}}$$

Suppose  $f$  is sampled at  $t_n$  with  $t_n \equiv n\Delta t$  and  $\Delta t \equiv \frac{T}{N}$ .

- $1/\Delta t$  is the **sample frequency**,
- $f_n \equiv f(t_n)$  are the **sampled function values**.

*We approximate  $\gamma_k$  with a Riemann integral using the sampled function values.*



## Fourier coefficients

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$$\tilde{\gamma}_k \equiv \frac{\Delta t}{T} \sum_{n=0}^{N-1} f(t_n) e^{-2\pi i t_n \frac{k}{T}} = \frac{1}{N} \sum_{n=0}^{N-1} f_n e^{-2\pi i \frac{nk}{N}}$$

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**Note.** The harmonic oscillations

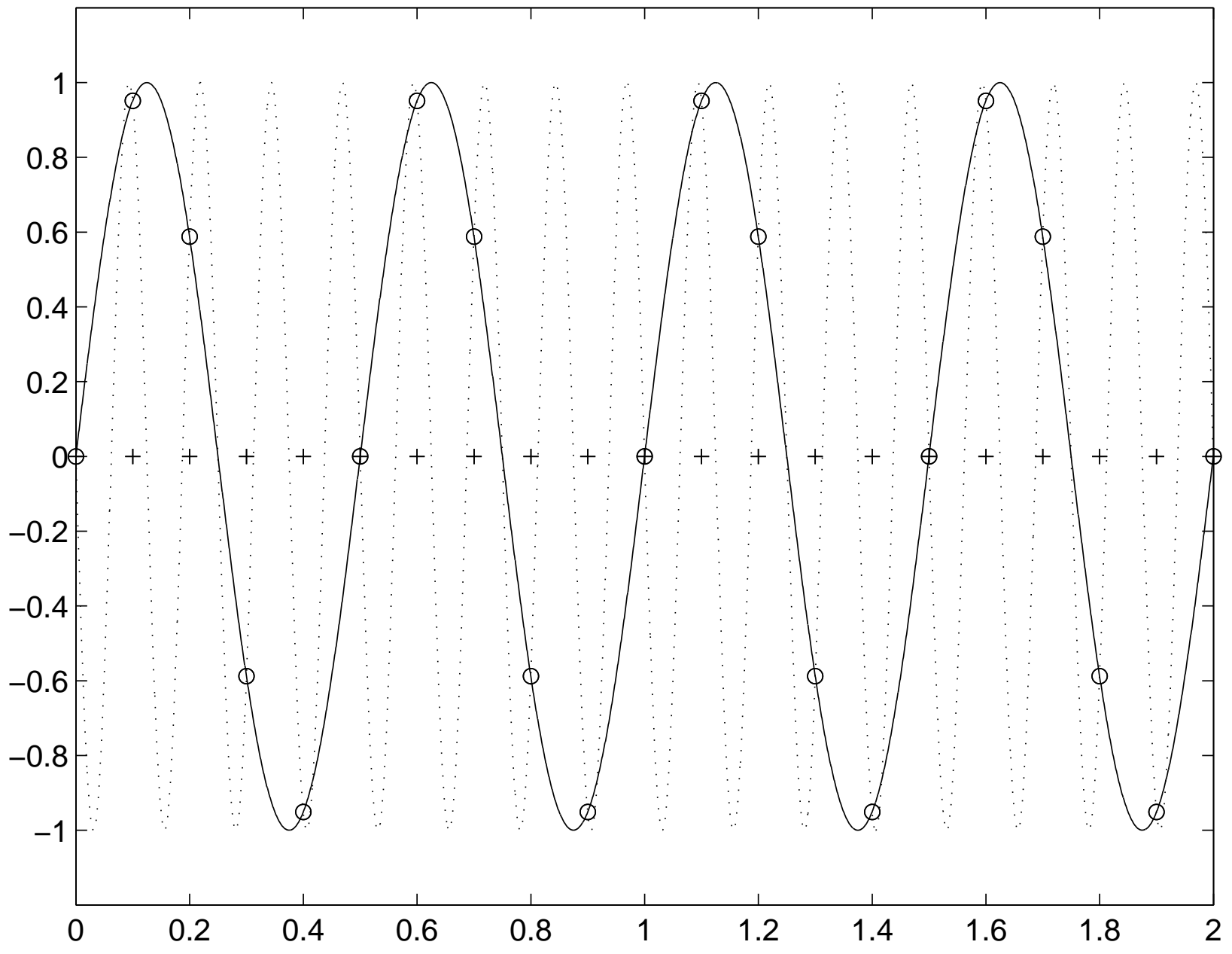
$$t \rightsquigarrow e^{2\pi i t \frac{k}{T}} \quad \text{and} \quad t \rightsquigarrow e^{2\pi i t \frac{k+N}{T}}$$

coincide at the sample points  $t_n$ .

The second oscillation is an **alias** of the first.

This phenomenon of **aliasing** has many consequences in discretised Fourier series.





## Fourier coefficients

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$$\tilde{\gamma}_k = \tilde{\gamma}_{k+jN} \quad (k, j \in \mathbb{Z}).$$

$$f_n = \sum_{k \in \mathbb{Z}} \mu_k e^{2\pi i \frac{nk}{N}}, \quad \text{where} \quad \mu_k \equiv \sum_{j \in \mathbb{Z}} \gamma_{k+jN}(f).$$

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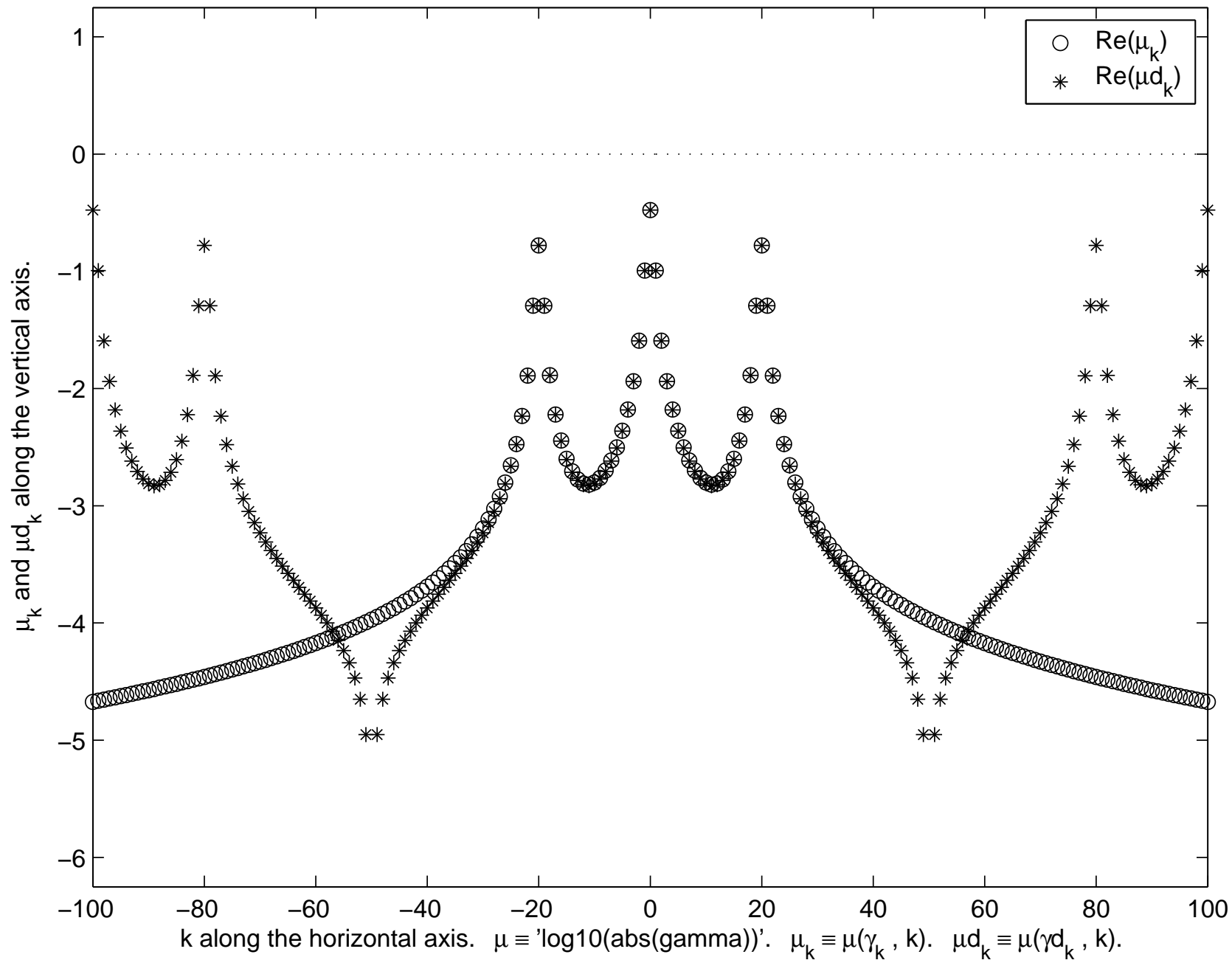
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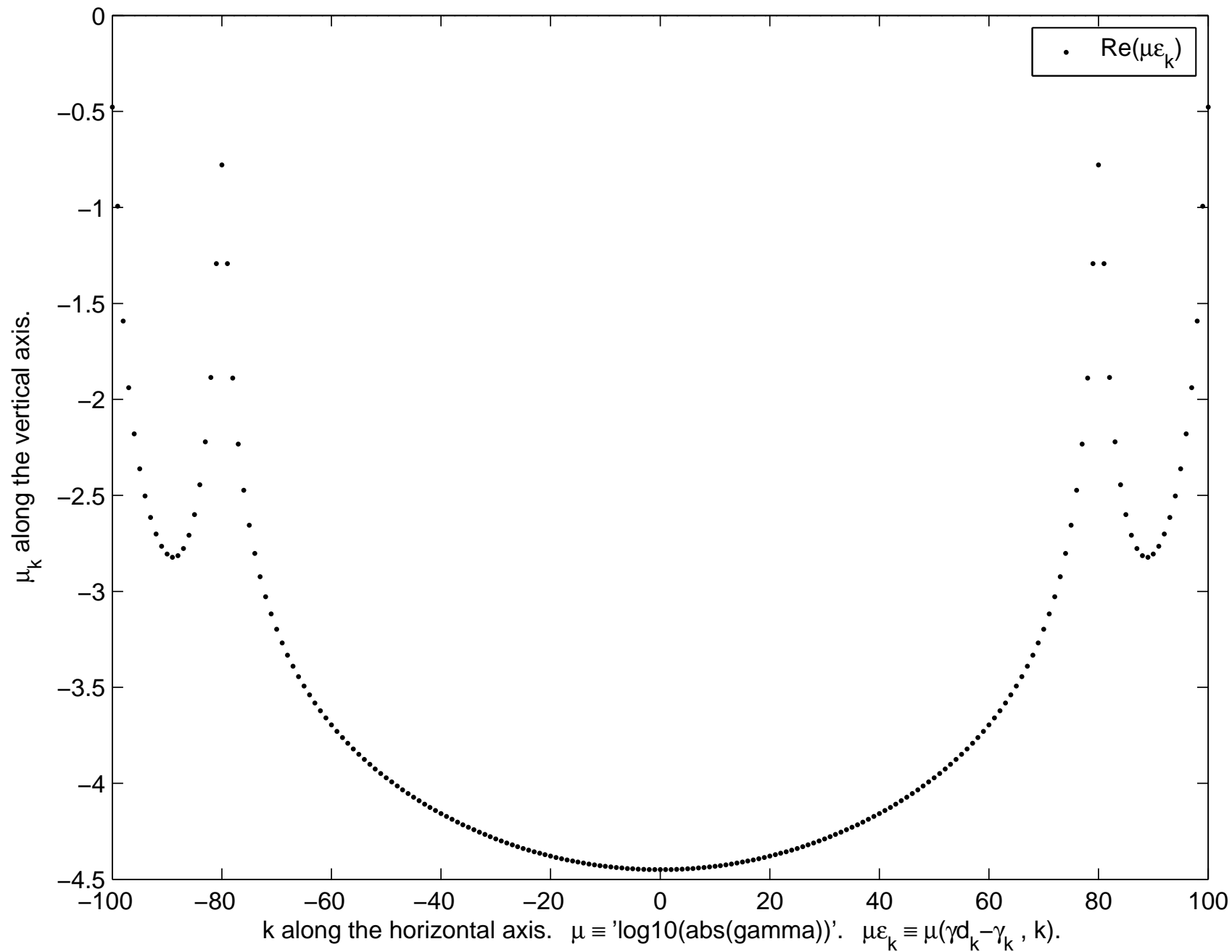
**Theorem.**  $\tilde{\gamma}_k = \mu_k = \gamma_k(f) + \sum_{j \neq 0} \gamma_{k+jN}(f)$ .

**Proof.** Apply next theorem.

The discrete Fourier coefficients  $\gamma d_k$  of  $f$  for  $N=100$ .



The error  $\varepsilon_k \equiv \gamma_k^d - \gamma_k$  in the discrete Fourier coefficients.



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# Discrete Fourier Transform

**Theorem.** Let  $(f_0, f_1, \dots, f_{N-1})$  be a sequence of complex numbers. Define the sequence  $(\tilde{\gamma}_0, \dots, \tilde{\gamma}_{N-1})$  by

$$\tilde{\gamma}_k \equiv \frac{1}{N} \sum_{n=0}^{N-1} f_n e^{-2\pi i \frac{nk}{N}} \quad (k = 0, \dots, N-1).$$

Then

$$f_n = \sum_{k=0}^{N-1} \tilde{\gamma}_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$



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**Note.** Except for the minus-sign in the exponential and the scaling  $\frac{1}{N}$  in the definition of the  $\tilde{\gamma}_k$ , the formulae are the same. *Some text books scale both formulae with  $\frac{1}{\sqrt{N}}$ .*

The sequence  $(\tilde{\gamma}_k)$  is the **Discrete Fourier Transform** of the sequence  $(f_n)$ . The theorem gives the inverse **DFT**.

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**Proof.** Let  $\ell(N)$  be the space of sequences  $\mathbf{f} \equiv (f_0, \dots, f_{N-1})$  of  $N$  complex numbers with inner product

$$\langle \mathbf{f}, \mathbf{g} \rangle \equiv \frac{1}{N} \sum_{n=0}^{N-1} f_n \overline{g_n} \quad (\mathbf{f}, \mathbf{g} \in \ell(N)).$$

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If  $k = m$ , then  $\zeta = 1$  and  $\langle \phi_k, \phi_k \rangle = 1$ .

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Note  $-N < k - m < N$ . Hence, if  $k - m \neq 0$ , then  $\zeta \neq 1$ .

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If  $k \neq m$ , then  $\zeta \neq 1$ ,  $\zeta^N = 1$ , and  $\langle \phi_k, \phi_m \rangle = \frac{1}{N} \frac{\zeta^N - 1}{\zeta - 1} = 0.$

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This proves orthonormality of the  $\phi_k$ , i.p., linear independence.

A dimension argument ( $\dim(\ell(N)) = N$ ) shows that the  $\phi_k$  form a basis.

# Discrete Fourier Transform

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The collection of  $\phi_k$  forms an orthonormal basis of  $\ell(N)$ .

In particular, 
$$\mathbf{f} = \sum_{k=0}^{N-1} \langle \mathbf{f}, \phi_k \rangle \phi_k$$

The def. of the inner product reveals that  $\tilde{\gamma}_k = \langle \mathbf{f}, \phi_k \rangle.$



# Discrete Fourier Transform

**Theorem.**  $\tilde{\gamma}_k \equiv \frac{1}{N} \sum_{n=0}^{N-1} f_n e^{-2\pi i \frac{nk}{N}} \Rightarrow f_n = \sum_{k=0}^{N-1} \tilde{\gamma}_k e^{2\pi i \frac{kn}{N}}.$

**Exercise.** For  $\tilde{\gamma} \equiv (\tilde{\gamma}_0, \dots, \tilde{\gamma}_{N-1}) \in \ell(N)$ , put

$$\mathcal{F}(\tilde{\gamma})_n \equiv \mathcal{F}_N(\tilde{\gamma})_n \equiv f_n \equiv \sum_{k=0}^{N-1} \tilde{\gamma}_k e^{2\pi i \frac{nk}{N}} \quad (k \in \mathbb{Z}).$$

With  $\mathbf{f} \equiv (f_0, \dots, f_{N-1})$ , prove that

$$\tilde{\gamma}_k = \frac{1}{N} \mathcal{F}(\mathbf{f})_{N-k} \quad (k = 0, \dots, N-1).$$

Note that the **DFT**  $\mathcal{F}_N$  produces  $N$ -periodic sequences.

**Conclusion.** The inverse **DFT** can easily be obtained from the **DFT** and visa versa.

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# Discrete Cosine Transform

The **DFT** requires complex arithmetic.

Moreover, as we know from Fourier series, the series converge slowly if the periodic function is discontinuous.

Any function  $f$  on a bounded interval, say,  $[0, T]$  can be extended to a  $T$ -periodic function. The obvious extension

$$f(t) \equiv f(t - jT) \quad (t \in \mathbb{R}, j \in \mathbb{Z}),$$

may lead to a discontinuous function on  $\mathbb{R}$  (if  $f(0) \neq f(T)$ ) with slowly decreasing Fourier coefficients  $\gamma_k(f)$ .

With the even extension first

$$f(t) \equiv f(-t), \quad f(t) \equiv f(t - 2jT) \quad (t \in \mathbb{R}, j \in \mathbb{Z})$$

we have an even  $2T$ -periodic function that is continuous whenever  $f$  is. In particular,  $\gamma_k$  of this function are real.

## Discrete Cosine Transform

Similarly, if  $\mathbf{f} \in \ell(N)$ , then complex arithmetic is avoided and at the same time faster decreasing discrete Fourier coefficients  $\tilde{\gamma}_k$  are obtained by extending  $\mathbf{f}$  first to an even function before extending to a periodic function.

For ease of notation, we put  $\gamma_k$  instead of  $\tilde{\gamma}_k$ .

## Discrete Cosine Transform

**Example.** Suppose  $\mathbf{f} = (f_0, \dots, f_N) \in \ell(N + 1)$ .

Extend  $\mathbf{f}$  to an function that is even (around  $n = N$ ):

$$\begin{aligned}\mathbf{g} &\equiv (f_0, f_1, \dots, f_{N-1}, f_N, f_{N-1}, \dots, f_2, f_1) \\ &= (g_0, g_1, \dots, g_{N-1}, g_N, g_{N+1}, \dots, g_{2N-2}, g_{2N-1})\end{aligned}$$

Note that the extension to a  $2N$ -periodic function is even also around  $n = 0$ .

The extended sequence is  $(\dots, \mathbf{g}, \mathbf{g}, \mathbf{g}, \dots)$ , to which we also shall refer to as  $\mathbf{g}$ .

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The **DFT** of  $\mathbf{g}$  is

$$\begin{aligned}\gamma_k &= \frac{1}{2N} \sum_{n=0}^{2N-1} g_n e^{-2\pi i \frac{kn}{2N}} \\ &= \frac{1}{2N} [f_0 + (-1)^k f_N] + \frac{1}{N} \sum_{n=1}^{N-1} f_n \cos(2\pi \frac{kn}{N})\end{aligned}$$

## Discrete Cosine Transform

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Note that, as  $\mathbf{g}$ ,  $(\gamma_k)$  is even around  $k = 0$  and  $k = N$ .

In particular,  $\gamma_k$  has to be computed for  $k = 0, \dots, N$  only.

## Discrete Cosine Transform

**Example.** Suppose  $\mathbf{f} = (f_0, \dots, f_N) \in \ell(N + 1)$ .

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Note that, as  $\mathbf{g}$ ,  $(\gamma_k)$  is even around  $k = 0$  and  $k = N$ .

Therefore, the inverse **DFT**, for  $n = 0, \dots, N$ , is

$$g_n = f_n = [\gamma_0 + (-1)^n \gamma_N] + 2 \sum_{k=1}^{N-1} \gamma_k \cos(2\pi \frac{kn}{N})$$



# Discrete Cosine Transform

There are a number of ways to extend a finite sequence to a sequence of length  $2N$  that is even.

**Example.** Suppose  $\mathbf{f} = (f_0, \dots, f_{N-1}) \in \ell(N)$ .

Then the extension

$$\mathbf{g} \equiv (\mathbf{f}, \mathbf{f}^T) \quad \text{with} \quad \mathbf{f}^T \equiv (f_{N-1}, f_{N-2}, \dots, f_1, f_0)$$

leads to an  $2N$ -periodic function  $\mathbf{g}$  that is even around  $n = -\frac{1}{2}$  and  $n = N - \frac{1}{2}$ .

This leads to the so-called **DCT-II** transform:

**DCT-II.** With  $\phi_{n,k} \equiv \cos\left(\pi\left(n + \frac{1}{2}\right)\frac{k}{N}\right)$ ,

$$\gamma_k = \frac{1}{N} \sum_{n=0}^{N-1} f_n \phi_{n,k}, \quad f_n = \gamma_0 + 2 \sum_{k=1}^{N-1} \gamma_k \phi_{n,k}$$

# Discrete Cosine Transform

There are a number of ways to extend a finite sequence to a sequence of length  $2N$  that is even. The first extension that we considered (even around 0 and  $N$ ) is called **DCT-I**, the second (even around  $-\frac{1}{2}$ ,  $N - \frac{1}{2}$ ) is **DCT-II**. The **DCT-II** seems to be the most popular one in practice and is often simply called **the DCT**.

Odd extensions lead to sines rather than cosines. However, sines are cosines up to some phase shift and with some simple manipulation, odd extensions also lead to transforms involving cosines only, to the so called **DCT-III** and **DCT-IV**. **DCT-IV** is the standard **DCT** in Matlab:

**DCT-IV**. With  $\phi_{n,k} \equiv \cos\left(\frac{\pi}{N}\left(n + \frac{1}{2}\right)\left(k + \frac{1}{2}\right)\right)$ ,

$$\gamma_k = \frac{1}{N} \sum_{n=0}^{N-1} f_n \phi_{n,k}, \quad f_n = 2 \sum_{k=0}^{N-1} \gamma_k \phi_{n,k}.$$

# Discrete Cosine Transform

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In the above **DCTs**, we extended to an even sequence **g** of length  $2N$ . Extension to an even sequence **g** of length  $2N - 1$  leads to **DCTs** of type **V**, **VI**, **VII** and **VIII**. These **DCT** seem to be rarely used in practice.

# Applications of DCT

- **Image compression.**

**Goal.** Compression.

2-dimensional (and 3-d) **DCT-II** is used with  $N$  low.

JPEG, MJPEG, MPEG use **DCT-II** on  $8 \times 8$  blocks

- **Audio compression.**

**Goal.** Compression and **spectral information**: the techniques in audio compression exploit psychological facts on how we hear combinations of harmonic oscillations, that is, compression depends on the distribution of frequencies.

A related transform, **Modified DCT**, is used in AAC, Vorbis, MP3.

- **Partial Differential Equations.** **DCTs** are used for solving PDEs, where the variants of **DCT** correspond to (slightly) different boundary conditions.

## Modified DCT

Let  $(\dots, f_1, f_0, \dots, f_{N-1}, f_N, \dots)$  be a long sequence of (sampled) function values. Then it is not feasible to compute the **FT**. As an alternative a part  $(f_0, \dots, f_{N-1})$  is considered. However **DFT** implicitly extends periodically. This will (probably) introduce 'jumps' in the function, implying slowly decreasing **DFT** coefficients.

Formally this argument does not apply to a discrete function, which, in some sense, has 'jumps' in all points  $t_n$ . But the argument **does** apply to a ( $T$ -periodic) function on  $\mathbb{R}$ . And discretization carries over the properties, in some approximate sense, to the discretized version.

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- The **DFT** coefficients may form bad approximations of the Fourier coefficients of interest,
- The compressibility properties may seriously deteriorate.

The even extension first, as is incorporated in **DCT**, weakens this effect. But, since the even extension, still introduces 'jumps' in the 'derivative', the effect is still noticeable. The odd extension (as in **DCT-IV**) may need special attention.

## Modified DCT

Let  $(\dots, f_1, f_0, \dots, f_{N-1}, f_N, \dots)$  be a long sequence of (sampled) function values.

'Discontinuities' at the ends are 'softened' by multiplying with a smooth function that is zero at the ends. For instance, with

$$\mathbf{S} \equiv (s_0, \dots, s_{2N-1}) \quad \text{with} \quad s_j \equiv \sin^2 \left( \frac{\pi}{2N} \left( j + \frac{1}{2} \right) \right),$$

apply the **DCT** to

$$(s_0 f_0, s_1 f_1, \dots, s_{2N-1} f_{2N-1}), \quad (s_0 f_{2N}, \dots, s_{2N-1} f_{4N-1}), \quad \dots$$

## Modified DCT

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To avoid loss of information, apply the **DCT** also to the middle parts, to

$$(s_0 f_N, s_1 f_{N+1}, \dots, s_{2N-1} f_{3N-1}), \quad (s_0 f_{3N}, \dots, s_{2N-1} f_{5N-1}), \quad \dots$$



## Modified DCT

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apply the **DCT** to

$$s_0 f_0, s_1 f_1, \dots, s_{2N-1} f_{2N-1} \quad , \quad s_0 f_{2N}, \dots, s_{2N-1} f_{4N-1} \quad , \quad \dots$$

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Note that the two sequences (when the parts are grouped in a long sequences ) add to the original sequence:

$$\sin^2 \phi + \cos^2 \phi = 1.$$

## Modified DCT

Let  $(\dots, f_1, f_0, \dots, f_{N-1}, f_N, \dots)$  be a long sequence of (sampled) function values.

'Discontinuities' at the ends are 'softened' by multiplying with a smooth function that is zero at the ends. For instance, with

$$\mathbf{S} \equiv (s_0, \dots, s_{2N-1}) \quad \text{with} \quad s_j \equiv \sin^2 \left( \frac{\pi}{2N} \left( j + \frac{1}{2} \right) \right),$$

apply the **DCT** to

$$s_0 f_0, s_1 f_1, \dots, s_{2N-1} f_{2N-1} \quad , \quad s_0 f_{2N}, \dots, s_{2N-1} f_{4N-1} \quad , \quad \dots$$

To avoid loss of information, apply the **DCT** also to the middle parts, to

$$s_0 f_N, s_1 f_{N+1}, \dots, s_{2N-1} f_{3N-1} \quad , \quad s_0 f_{3N}, \dots, s_{2N-1} f_{5N-1} \quad , \quad \dots$$

Note that the two sequences (when the parts are grouped in a long sequences) add to the original sequence: **MDCT** is designed to deal with this overlapping type of grouping.

## Modified DCT

Let  $(\dots, f_1, f_0, \dots, f_{N-1}, f_N, \dots)$  be a long sequence of (sampled) function values. Assume  $M \equiv N/2$  is an integer.

With  $\phi_{n,k} \equiv \cos\left(\frac{\pi}{N}\left(k + \frac{1}{2}\right)\left(n + M + \frac{1}{2}\right)\right)$ , we have

$$\gamma_k = \frac{1}{N} \sum_{n=0}^{2N-1} f_n \phi_{n,k} \quad (k = 0, \dots, N-1), \quad \tilde{f}_n = \sum_{k=0}^{N-1} \gamma_k \phi_{n,k} \quad (n = 0, \dots, 2N-1)$$

Using sequences of length  $N$ , we summarize this as

$$(F_1, F_2) \rightsquigarrow \Gamma_1, \quad \Gamma_1 \rightsquigarrow (\tilde{F}_1, \tilde{G}_2).$$

**Theorem.**  $(\tilde{F}_1, \tilde{G}_2) = (F_1 - F_1^\top, F_2 + F_2^\top)$ .

**Proof.** MDCT is a form of DCT-IV.

## Modified DCT

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**Modified DCT** applies this to long sequences as

$$\begin{aligned} (F_1, F_2, F_3, F_4, \dots, F_k) &\rightsquigarrow (\Gamma_1, \Gamma_2, \Gamma_3, \dots, \Gamma_{k-1}) \\ &\rightsquigarrow (\tilde{F}_1, \tilde{G}_2 + \tilde{F}_2, \tilde{G}_3 + \tilde{F}_3, \tilde{G}_4 + \tilde{F}_4, \dots, \tilde{G}_{k-1} + \tilde{F}_{k-1}, \tilde{G}_k) \end{aligned}$$

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Note that we have exact reconstruction except for the first and last block (usual choice:  $F_1 = \mathbf{0}$ ,  $F_k = \mathbf{0}$ ).

## Modified DCT

Let  $(\dots, f_1, f_0, \dots, f_{N-1}, f_N, \dots)$  be a long sequence of (sampled) function values. Assume  $M \equiv N/2$  is an integer.

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Including windows, with  $C = S^\top$ ,  $\mathbf{S} = (S, C)$ ,

$$(F_1, F_2) \rightsquigarrow (SF_1, CF_2) \rightsquigarrow \tilde{\Gamma}_1 \rightsquigarrow (\tilde{F}_1, \tilde{G}_2) \rightsquigarrow (S\tilde{F}_1, C\tilde{G}_2).$$

### Theorem.

$$(S\tilde{F}_1, C\tilde{G}_2) = (S^2 F_1 - SCF_1^\top, C^2 F_2 + SCF_2^\top).$$

With  $S^2 + C^2 = 1$ , application to long sequences leads to exact reconstruction.



# Program

- Computing Fourier Coefficients
- Discrete Fourier Transform
- Discrete Cosine Transform
- Fast Fourier Transform
- Computing Fourier Integrals

# Fast Fourier Transform

Suppose the sequence  $\gamma \equiv (\gamma_0, \dots, \gamma_{N-1}) \in \ell(N)$  is available. The naive way of computing the **DFT**

$$\mathcal{F}(\gamma)_n \equiv f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

requires more than  $2N^2$  **floating point operations** (additions, multiplications): for each of the  $N$   $n$ s,  $2N$  **flop**.

In practice  $N$  is huge.

$N$  of the order of  $10^6 \sim 10^8$  is not exceptional.

*Gauss [first half of the 19th century], Runge [1903] and Cooley & Tukey [1965] in the most cited mathematical paper ever, proposed a computational scheme, **FFT**, that reduces the computational costs to  $2N \log_2(N)$  **flop**. For, e.g.,  $n = 2^{20} \approx 10^6$ , this makes a difference with  $2N^2$  of 1 sec versus 3:30 hours.*

# Fast Fourier Transform

Suppose  $N = 2^\ell$  for some  $\ell \in \mathbb{N}$ . Put  $M = 2^{\ell-1} = \frac{1}{2}N$ .

$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

We split the sum into one with even indices and one with odd indices.

# Fast Fourier Transform

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$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

$$f_n = \left( \sum_{2k < N} \gamma_{2k} e^{2\pi i \frac{kn}{M}} \right) + \left( \sum_{2k+1 < N} \gamma_{2k+1} e^{2\pi i \frac{kn}{M}} \right) e^{2\pi i \frac{kn}{N}}.$$

or

$$f_n = f_{e,n} + f_{o,n} e^{2\pi i \frac{kn}{N}} \quad \text{with} \quad f_{e,n} \equiv \sum_{2k < N} \gamma_{2k} e^{2\pi i \frac{kn}{M}}$$

and  $f_{o,n}$  defined similarly:

$$f_{e,n} \equiv \sum_{2k+1 < N} \gamma_{2k+1} e^{2\pi i \frac{kn}{M}}.$$

# Fast Fourier Transform

Suppose  $N = 2^\ell$  for some  $\ell \in \mathbb{N}$ . Put  $M = 2^{\ell-1} = \frac{1}{2}N$ .

$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

$$f_n = f_{e,n} + f_{o,n} e^{2\pi i \frac{n}{N}} \quad \text{with} \quad f_{e,n} \equiv \sum_{2k < N} \gamma_{2k} e^{2\pi i \frac{kn}{M}}$$

Note that  $f_{e,n} = f_{e,n+M}$ . Similarly,  $f_{o,n} = f_{o,n+M}$ .

Moreover,  $\exp(\pi i \frac{n+M}{M}) = -\exp(\pi i \frac{n}{M})$ . Therefore,

$$\begin{aligned} f_n &= f_{e,n} + f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1), \\ f_{n+M} &= f_{e,n} - f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1). \end{aligned}$$

# Fast Fourier Transform

Suppose  $N = 2^\ell$  for some  $\ell \in \mathbb{N}$ . Put  $M = 2^{\ell-1} = \frac{1}{2}N$ .

$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

$$f_n = f_{e,n} + f_{o,n} e^{2\pi i \frac{n}{N}} \quad \text{with} \quad f_{e,n} \equiv \sum_{2k < N} \gamma_{2k} e^{2\pi i \frac{kn}{M}}$$

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To compute  $f_n$  for all  $n = 0, \dots, N-1$ , we need to compute two Fourier transforms  $f_{e,n}$  and  $f_{o,n}$  for sequences of coefficients of half of the length and for only half of the number of  $ns$  ( $n = 0, \dots, M-1$ ).

# Fast Fourier Transform

Suppose  $N = 2^\ell$  for some  $\ell \in \mathbb{N}$ . Put  $M = 2^{\ell-1} = \frac{1}{2}N$ .

$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

With  $f_{e,n} \equiv \sum_{2k < N} \gamma_{2k} e^{2\pi i \frac{kn}{M}}$ ,  $f_{o,n} \equiv \sum_{2k+1 < N} \gamma_{2k+1} e^{2\pi i \frac{kn}{M}}$ ,

we have  $f_n = f_{e,n} + f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1),$

$$f_{n+M} = f_{e,n} - f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1).$$

# Fast Fourier Transform

Suppose  $N = 2^\ell$  for some  $\ell \in \mathbb{N}$ . Put  $M = 2^{\ell-1} = \frac{1}{2}N$ .

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we have

$$f_n = f_{e,n} + f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1),$$
$$f_{n+M} = f_{e,n} - f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1).$$

Let  $\kappa_\ell$  be the number of **flop** required to compute the **DFT** of length  $M = 2^\ell$ . Then, the above implies that

$$\kappa_\ell = 2\kappa_{\ell-1} + 1.5N.$$

We need  $N$  additions (subtractions),  $M$  multiplications;  
For now, we neglected the costs for computing  $e^{\pi i \frac{n}{M}}$ .



# Fast Fourier Transform

Suppose  $N = 2^\ell$  for some  $\ell \in \mathbb{N}$ . Put  $M = 2^{\ell-1} = \frac{1}{2}N$ .

$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

With  $f_{e,n} \equiv \sum_{2k < N} \gamma_{2k} e^{2\pi i \frac{kn}{M}}$ ,  $f_{o,n} \equiv \sum_{2k+1 < N} \gamma_{2k+1} e^{2\pi i \frac{kn}{M}}$ ,

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Let  $\kappa_\ell$  be the number of **flop** required to compute the **DFT** of length  $M = 2^\ell$ . Then, the above implies that

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We need  $N$  additions (subtractions),  $M$  multiplications; For now, we neglected the costs for computing  $e^{\pi i \frac{n}{M}}$ .

We can repeat the partitioning trick to  $f_{e,n}$  and  $f_{o,n}$ .

# Fast Fourier Transform

Suppose  $N = 2^\ell$  for some  $\ell \in \mathbb{N}$ . Put  $M = 2^{\ell-1} = \frac{1}{2}N$ .

$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

With  $f_{e,n} \equiv \sum_{2k < N} \gamma_{2k} e^{2\pi i \frac{kn}{M}}$ ,  $f_{o,n} \equiv \sum_{2k+1 < N} \gamma_{2k+1} e^{2\pi i \frac{kn}{M}}$ ,

we have

$$f_n = f_{e,n} + f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1),$$
$$f_{n+M} = f_{e,n} - f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1).$$

Let  $\kappa_\ell$  be the number of **flop** required to compute the **DFT** of length  $M = 2^\ell$ . Then, the above implies that

$$\kappa_\ell = 2\kappa_{\ell-1} + 1.5N = 2(2\kappa_{\ell-2} + 1.5M) + 1.5N.$$

# Fast Fourier Transform

Suppose  $N = 2^\ell$  for some  $\ell \in \mathbb{N}$ . Put  $M = 2^{\ell-1} = \frac{1}{2}N$ .

$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

With  $f_{e,n} \equiv \sum_{2k < N} \gamma_{2k} e^{2\pi i \frac{kn}{M}}$ ,  $f_{o,n} \equiv \sum_{2k+1 < N} \gamma_{2k+1} e^{2\pi i \frac{kn}{M}}$ ,

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Let  $\kappa_\ell$  be the number of **flop** required to compute the **DFT** of length  $M = 2^\ell$ . Then, the above implies that

$$\kappa_\ell = 2\kappa_{\ell-1} + 1.5N = 4\kappa_{\ell-2} + 2 \cdot 1.5N.$$

Repeating, the partitioning trick to  $f_{e,n}$  and  $f_{o,n}$ , etc., and using the fact that  $\kappa_0 = 0$ , shows that

$$\kappa_\ell = 2\kappa_{\ell-1} + 1.5N = 1.5\ell N.$$

# Fast Fourier Transform

Suppose  $N = 2^\ell$  for some  $\ell \in \mathbb{N}$ . Put  $M = 2^{\ell-1} = \frac{1}{2}N$ .

$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

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Repeating this partitioning trick recursively down to level  $\ell = 0$  is **Fast Fourier Transform**.

**Theorem.** FFT requires  $(1.5\ell + 0.5)N$  flop.

# Fast Fourier Transform

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$$\begin{aligned} \text{we have} \quad f_n &= f_{e,n} + f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1), \\ f_{n+M} &= f_{e,n} - f_{o,n} e^{\pi i \frac{n}{M}} \quad (n = 0, \dots, M-1). \end{aligned}$$

Repeating this partitioning trick recursively down to level  $\ell = 0$  is **Fast Fourier Transform**.

**Theorem.** FFT requires  $(1.5\ell + 0.5)N$  flop.

**Proof.** The  $0.5N$  comes from the computation of  $e^{\pi i \frac{n}{M}}$ , which can be computed as  $\zeta^n = \zeta^{n-1} \zeta$  with  $\zeta = e^{\pi i \frac{1}{M}}$ . Note that,  $e^{\pi i n 2^{-\ell+j}} = e^{\pi i (2^j n) 2^{-\ell}}$ .

## DFT & FFT as matrix multiplication

Let  $\mathbf{f} = (f_0, \dots, f_{N-1})^T$  and  $\boldsymbol{\gamma} = (\gamma_0, \dots, \gamma_{N-1})^T$  be such that

$$f_n = \sum_{k=0}^{N-1} \gamma_k e^{2\pi i \frac{kn}{N}} \quad (n = 0, \dots, N-1).$$

We will represent the **DFT** and **FFT** with respect to column vectors here:  $\cdot^T$  is the transpose for vectors.

Note that, for consistency of notation, we selected the first index of the vectors to be 0 rather than 1.

Our matrices will also be indexed from 0 on: the left top matrix element will be the (0,0)-entry.

## DFT & FFT as matrix multiplication

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Let  $\mathbf{F}$  be the  $N \times N$  matrix with  $(n, k)$ -entry  $e^{2\pi i \frac{kn}{N}}$ . Then

$$\mathbf{f} = \mathbf{F}\boldsymbol{\gamma}$$

This represents the **DFT** (Discrete Fourier transform) as a **MV** (matrix-vector multiplication).

- Note that  $\frac{1}{N} \mathbf{F}^* \mathbf{F} = \mathbf{I}$ : except for the scaling  $N$ , the **DFT** matrix  $\mathbf{F}$  is unitary. Here  $\mathbf{F}^* \equiv \overline{\mathbf{F}}^T$ .
- The inverse  $\frac{1}{N} \mathbf{F}^*$  of  $\mathbf{F}$  represents the inverse **DFT**.
- The **DFT** matrix  $\mathbf{F}$  is full: none of its entries is zero. This makes the **MV** expensive.

## DFT & FFT as matrix multiplication

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**FFT.** Now suppose  $N = 2^\ell$ . Put  $M \equiv 2^{\ell-1}$ .

Let  $\mathbf{F}_\ell \equiv \mathbf{F}$  be the **DFT** for level  $\ell$ , i.e., for  $N = 2^\ell$ .

The first step in **FFT** can be written as

$$\mathbf{f} = \begin{bmatrix} \mathbf{f}' \\ \mathbf{f}'' \end{bmatrix} = \begin{bmatrix} \mathbf{I}_{\ell-1} & +\mathbf{D}_{\ell-1} \\ \mathbf{I}_{\ell-1} & -\mathbf{D}_{\ell-1} \end{bmatrix} \begin{bmatrix} \mathbf{f}_e \\ \mathbf{f}_o \end{bmatrix}.$$

Here,  $\mathbf{f}' \equiv (f_0, \dots, f_{M-1})^T$ ,  $\mathbf{f}'' \equiv (f_M, \dots, f_{N-1})^T$ ,

$\mathbf{I}_{\ell-1}$  is the  $M \times M$  identity matrix,

$\mathbf{D}_{\ell-1}$  is the  $M \times M$  diagonal matrix with  $(n, n)$ -entry  $e^{\pi i \frac{n}{M}}$ .

Note that the even diagonal entries of  $D_{\ell-1}$  form  $D_{\ell-2}, \dots$



## DFT & FFT as matrix multiplication

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where  $\boldsymbol{\gamma}_e = (\gamma_0, \gamma_2, \dots)^T$ ,  $\boldsymbol{\gamma}_o = (\gamma_1, \gamma_3, \dots)^T$ .

## DFT & FFT as matrix multiplication

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where  $P$  is the permutation that puts the even indexed  $\gamma_k$  first: following Matlab  $\gamma(P) \equiv (\gamma_{P(0)}, \dots, \gamma_{P(N-1)})^T$ .

$P(k) \equiv 2k$  ( $k < M$ ),  $P(k) \equiv 2(k - M) + 1$  ( $k > M$ ).

## DFT & FFT as matrix multiplication

Let  $\mathbf{f} = (f_0, \dots, f_{N-1})^T$  and  $\gamma = (\gamma_0, \dots, \gamma_{N-1})^T$  be such that

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This decomposes  $\mathbf{F}$  as a product of a very sparse unitary matrix (only 2 non-zeros per row), two **DFTs** of half size and a permutation. Repeating this decomposition gives **FFT**:  $\mathbf{F}$  is a product of only  $\ell$  very sparse unitary matrices and a permutation.

## FFT for sequences of any length?

Suppose  $N = q^\ell$  for some  $q \in \mathbb{N}, q > 2$ .

Then we can design a **FFT** algorithm similar to the one for  $q = 2$ . For instance, if  $q = 3$ , and  $(\gamma_0, \dots, \gamma_{N-1})$  is a sequence of length  $N$ , then we can group the coefficients in three classes  $(\gamma_{3k})$ ,  $(\gamma_{3k+1})$  and  $(\gamma_{3k+2})$  instead of the two as for  $q = 2$  (the one with even indices and one with odd indices) and we can decompose the  $f_n$  accordingly.

$q$  is the **radix** of the **FFT**.

The computational costs are in the order of  $N \log_3 N$ :

Comp. Costs  $\approx C_q N \log_3 N$  for some  $C_q > 0$ .

**Property.** **FFT** with radix 4 allows the most efficient implementation (i.e.,  $4 = \operatorname{argmin}_q C_q N \log_q N$ ).

## FFT for sequences of any length?

If, say,  $N = 2^{\ell_1} 3^{\ell_2}$ , then we can form a **FFT** by

- applying the **FFT** with radix 2  $\ell_1$ -times
- followed by  $\ell_2$ -times the **FFT** with radix 3.

More generally,

## FFT for sequences of any length?

We can factorise any  $N \in \mathbb{N}$ , that is, we can decompose any  $N$  into a product of prime factors and we can design a **FFT** for sequences of length  $N$  that is a mixture of **FFTs** of radix  $p_j$  with  $p_j$  the primes that occur in the factors.

However, computationally, this approach is not attractive:

- we have to factorise  $N$
- coding of such a **FFT** with a mixture of **FFTs** with different radices is messy
- if the primes are large (with the extremal situation where  $N$  itself is prime), then the **FFT** is not faster.

## FFT for sequences of any length?

If, for instance, we have to compute Fourier coefficients of a  $T$ -periodic function  $f : \mathbb{R} \rightarrow \mathbb{C}$ , then we can select the sample frequency  $1/\Delta t$  as we like (with the only restriction that it is sufficiently large), for instance,

$$\Delta t = T/N \quad \text{with} \quad N = 2^\ell.$$

### Conclusion.

Some application allow to select  $N$  to be a power of 2.

## FFT for sequences of any length?

Some applications allow sequences  $(\gamma_0, \dots, \gamma_{M-1})$  of length  $M$  to be extended to sequences of length  $2^\ell$  (with  $\ell$  such that  $2^{\ell-1} < M \leq 2^\ell$ ) by appending with zeros.

**Example.** The **convolution product**  $\alpha \star \beta$  of the sequence  $\alpha = (\alpha_0, \dots, \alpha_{M-1})$  and  $\beta = (\beta_0, \dots, \beta_{M-1})$  is defined by

$$(\alpha \star \beta)_k \equiv \sum_j \alpha_j \beta_{k-j} \quad (k = 0, \dots, 2M - 2),$$

where we sum over all  $j \in \mathbb{Z}$  for which  $\alpha_j$  and  $\beta_{k-j}$  exists, that is,  $j$  such that  $j, k - j \in \{0, \dots, M - 1\}$ .

### Application.

If  $p$  is the polynomial  $p(x) = \alpha_0 + \alpha_1 x + \dots + \alpha_{M-1} x^{M-1}$  and  $q(x) = \beta_0 + \beta_1 x + \dots + \beta_{M-1} x^{M-1}$ , then  $(\alpha \star \beta)_k$  are the coefficients of the product polynomial  $pq$ .



## FFT for sequences of any length?

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Note that the value of  $\alpha \star \beta$  does not change if we extend  $\alpha$  and  $\beta$  by appending with zeros. Therefore, we may assume that the length of  $\alpha$  and  $\beta$  is  $M = 2^\ell$ .

## FFT for sequences of any length?

Some applications allow sequences  $(\gamma_0, \dots, \gamma_{M-1})$  of length  $M$  to be extended to sequences of length  $2^\ell$  (with  $\ell$  such that  $2^{\ell-1} < M \leq 2^\ell$ ) by appending with zeros.

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$$(\alpha \star \beta)_k \equiv \sum_j \alpha_j \beta_{k-j} \quad (k = 0, \dots, 2M - 2).$$

Assume that the length of  $\alpha$  and  $\beta$  is  $M = 2^\ell$ . Append  $\alpha$  and  $\beta$  with zeros to sequences of length  $N \equiv 2^{\ell+1}$ . Next, extend  $\alpha$  and  $\beta$  periodically (period  $N$ ) and define  $\star_N$ :

$$(\alpha \star_N \beta)_k \equiv \sum_{j=0}^{N-1} \alpha_j \beta_{k-j} \quad (k = 0, \dots, N - 1).$$

Note that the definitions of  $\alpha \star \beta$  are consistent (lead to the same values for  $k < N$ ).

# Discrete Convolution Products

**Definition.** For  $\alpha, \beta \in \ell(N)$ , let

$$(\alpha \star_N \beta)_k \equiv \sum_{j=0}^{N-1} \alpha_j \beta_{k-j} \quad (k = 0, \dots, N-1),$$

where  $\beta_{k-j} \equiv \beta_{N+k-j}$  if  $k-j < 0$  (periodic extension).

**Theorem.**  $\mathcal{F}_N(\alpha \star_N \beta) = \mathcal{F}_N(\alpha) \cdot \mathcal{F}_N(\beta)$ , where the  $\cdot$ -product is coordinate wise (the **Hadamard product**).

Suppose  $2^{\ell-1} < N < 2^\ell$ . Put  $L \equiv 2 \cdot 2^\ell$ .

Form  $\tilde{\beta} \equiv (\beta, \mathbf{0}, \beta)$  to a sequence of length  $L$ .

Form  $\tilde{\alpha} \equiv (\alpha, \mathbf{0}, \mathbf{0})$  to a sequence of length  $L$ .

**Property.**  $(\alpha \star_N \beta)_k = (\tilde{\alpha} \star_L \tilde{\beta})_k$  for  $k = 0, \dots, N-1$ .

**Corollary.**  $(\alpha \star_N \beta)_k = (\mathcal{F}_L^{-1}[\mathcal{F}_L(\tilde{\alpha}) \cdot \mathcal{F}_L(\tilde{\beta})])_k \quad (k < N)$ .

$\alpha \star_L \beta$  can be computed with three **DFT** of radix 2 plus  $L$  mult.. Costs:  $\leq 24N(\ell + 2)$  **flop** rather than  $0.5 N^2$ .

## FFT for sequences of any length?

Some applications allow sequences  $(\gamma_0, \dots, \gamma_{M-1})$  of length  $M$  to be extended to sequences of length  $2^\ell$  (with  $\ell$  such that  $2^{\ell-1} < M \leq 2^\ell$ ) by appending with zeros.

**Example.** The **convolution product**  $\alpha \star \beta$  of the sequence  $\alpha = (\alpha_0, \dots, \alpha_{M-1})$  and  $\beta = (\beta_0, \dots, \beta_{M-1})$  is defined by

$$(\alpha \star \beta)_k \equiv \sum_j \alpha_j \beta_{k-j} \quad (k = 0, \dots, 2M - 2).$$

**Conclusion.** In these applications the FFT is nothing more than an efficient computational tool. The quantities to be computed are in same domain as the inputs (time-domain rather than in frequency domain).

## Appending with zeros

Consider  $\gamma = (\gamma_0, \dots, \gamma_{M-1}) \in \ell(M)$  with  $2^{\ell-1} < M < N \equiv 2^\ell$ .  
Append  $\gamma$  with zeros to a sequence  $\gamma^+$  of length  $N$ :

$$\gamma^+ \equiv (\gamma_0, \dots, \gamma_{M-1}, 0, \dots, 0).$$

Observe that  $\mathcal{F}_M(\gamma) \neq \mathcal{F}_N(\gamma^+)$ , because

$$\sum_{k=0}^{M-1} \gamma_k e^{2\pi i \frac{kn}{M}} = \sum_{k=0}^{N-1} \gamma_k^+ e^{2\pi i \frac{kn}{M}} \neq \sum_{k=0}^{N-1} \gamma_k^+ e^{2\pi i \frac{kn}{N}}.$$

**Heuristics.** Appending with zeros introduces ‘discontinuities’, thus quickly decreasing Fourier coefficients maybe changed into slowly decreasing ones.

## Appending with zeros

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**Conclusion.** If the quantities of interest are in the ‘dual’ domain (frequency rather than time, or time rather than frequency), then appending zeros is not allowed.

## FFT for sequences of any length?

Consider  $\gamma = (\gamma_0, \dots, \gamma_{M-1}) \in \ell(M)$  with  $2^{\ell-1} < M < N \equiv 2^\ell$ .  
Put

$$\beta_k \equiv e^{-\pi i \frac{k^2}{M}} \quad (k = 0, \dots, M-1).$$

Note that  $e^{2\pi i \frac{kn}{M}} = \bar{\beta}_k \beta_{n-k} \bar{\beta}_n$ . Hence,

$$\mathcal{F}_M(\gamma)_n = \sum_{k=0}^{M-1} \gamma_k e^{2\pi i \frac{kn}{M}} = \left( \sum_{k=0}^{M-1} (\gamma_k \bar{\beta}_k) \beta_{n-k} \right) \bar{\beta}_n.$$

Apparently,

$$\mathcal{F}_M(\gamma) = ((\gamma \bar{\beta}) \star_M \beta) \bar{\beta}.$$

Here the convolution product is defined for  $M$ -periodic sequences.



## FFT for sequences of any length?

Consider  $\gamma = (\gamma_0, \dots, \gamma_{M-1}) \in \ell(M)$  with  $2^{\ell-1} < M < N \equiv 2^\ell$ .

**Property.** With  $\beta_k \equiv e^{-\pi i \frac{k^2}{M}}$ , we have that

$$\mathcal{F}_M(\gamma) = \mu \bar{\beta} \quad \text{with} \quad \mu \equiv (\gamma \bar{\beta}) \star_M \beta$$

As we saw before, the convolution product can be computed with three **DFT** of radix 2 (and length  $L \equiv 2N$ ), plus  $L$  multiplications. The multiplications  $\gamma \bar{\beta}$  and  $\mu \bar{\beta}$  require an additional  $2M$  multiplications.

# Program

- Computing Fourier Coefficients
- Discrete Fourier Transform
- Discrete Cosine Transform
- Fast Fourier Transform
- Computing Fourier Integrals

## Computing Fourier integrals

$f$  sampled at  $t_n = t_0 + n \Delta t$ .  $1/\Delta t$  sample frequency.  
For ease of notation, take  $t_0 = 0$  (otherwise shift by  $t_0$ ).

$$\hat{f}(\omega) \approx \int_{t_0}^{t_0+T} f(t) e^{-2\pi i t \omega} dt \approx \Delta t \sum_{n=0}^{N-1} f_n e^{-2\pi i n \Delta t \omega}$$

Here,  $T = N\Delta t$  and  $f_n = f(t_n)$ .

Of interest for  $\omega = \frac{k}{T}$  ( $k = 0, \dots, N - 1$ ).

$\hat{f}(\omega)$  to be computed by **DFT**.

Two 'discretizations'! **How accurate is this?**

$$\hat{f}(\omega) \approx \int_{t_0}^{t_0+T} f(t) e^{-2\pi i t \omega} dt$$

---

If  $f \in L^1(\mathbb{R})$  then,

for each  $\varepsilon > 0$ , there is a  $t_0$  and a  $T > 0$  such that

$$\int_{-\infty}^{t_0} |f(t)| dt < \varepsilon \quad \text{and} \quad \int_{t_0+T}^{\infty} |f(t)| dt < \varepsilon$$

However,

this observation is often only of theoretical interest.

In practice  $T$  can be large and huge values of  $N$  may be required, or spectral information is requested before all relevant function values  $f$  are available.

# Windowing

$$\hat{f}(\omega) \approx \int_{t_0}^{t_0+T} f(t) e^{-2\pi i t \omega} dt$$

Actually, we are computing the Fourier transform of

$$fW_{t_0}, \text{ where } W(t) = 1 \text{ if } 1 \leq t \leq T, \text{ and} \\ W(t) = 0 \text{ elsewhere}$$

and  $W_{t_0}(t) \equiv W(t - t_0)$ .

$W$  is a **time-window**.

Of interest: the difference between  $\hat{f}(\omega)$  and  $(\widehat{fW_{t_0}})(\omega)$ .

$\Phi(t, \omega) \equiv (\widehat{fW_t})(\omega)$  is called a **spectrogram** of  $f$ .

# Effects of windowing

## Computing Fourier integrals

$f$  sampled at  $t_n = t_0 + n \Delta t$ .  $1/\Delta t$  sample frequency.  
For ease of notation, take  $t_0 = 0$  (otherwise shift by  $t_0$ ).

$$\hat{f}(\omega) \approx \int_{t_0}^{t_0+T} f(t) e^{-2\pi i t \omega} dt \approx \Delta t \sum_{n=0}^{N-1} f_n e^{-2\pi i n \Delta t \omega}$$

Two 'discretizations'! How accurate is this?

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**Analysis.** Other order: first discretize, then window.



# Discretization

$$F(\omega) \equiv \Delta t \sum_{n=-\infty}^{\infty} f_n e^{-2\pi i n \Delta t \omega}$$

Relation  $\hat{f}(\omega)$  and  $F(\omega)$ ? Does this depend on  $\omega$ ?

$B_f \equiv \{\omega \in \mathbb{R} \mid |\hat{f}(\omega)| \neq 0\}$  is the **frequency band** of  $f$ .  
 $f$  is of **bounded bandwidth** if  $B_f \subset [-\Omega, +\Omega]$   
for some  $\Omega > 0$ : smallest  $\Omega$  is the **bandwidth**.

Suppose  $f$  is of bandwidth  $\leq \Omega$ .

$$f(t) = \sum_{k=-\infty}^{\infty} \gamma_k e^{2\pi i \frac{t}{T} k} \Leftrightarrow \gamma_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-2\pi i \frac{t}{T} k} dt$$

---

Take  $\Delta t = \frac{1}{2\Omega}$ , change  $-t \leftrightarrow \omega$ ,  $T \leftrightarrow 2\Omega$ ,  $n \leftrightarrow k \dots$

$$F(\omega) \equiv \Delta t \sum_{n=-\infty}^{\infty} f_n e^{-2\pi i n \Delta t \omega} \Leftrightarrow f_n = \int_{-\Omega}^{\Omega} F(\omega) e^{2\pi i \frac{\omega}{2\Omega} n} d\omega$$

$f$  of bandwidth  $\leq \Omega \Rightarrow$

$$f(t) = \int_{-\infty}^{\infty} \hat{f}(\omega) e^{2\pi i t \omega} d\omega = \int_{-\Omega}^{\Omega} \hat{f}(\omega) e^{2\pi i t \omega} d\omega$$

In particular,  $f_n = f(t_n) = \int_{-\Omega}^{\Omega} \hat{f}(\omega) e^{2\pi i \frac{n\omega}{2\Omega}} d\omega.$

$$f(t) = \sum_{k=-\infty}^{\infty} \gamma_k e^{2\pi i \frac{t}{T} k} \Leftrightarrow \gamma_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-2\pi i \frac{t}{T} k} dt$$

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$f$  of bandwidth  $\leq \Omega \Rightarrow$

$$\int_{-\Omega}^{\Omega} [\hat{f}(\omega) - F(\omega)] e^{2\pi i \frac{n\omega}{2\Omega}} d\omega = 0 \quad \forall n \in \mathbb{Z}.$$

$$f(t) = \sum_{k=-\infty}^{\infty} \gamma_k e^{2\pi i \frac{t}{T} k} \Leftrightarrow \gamma_k = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-2\pi i \frac{t}{T} k} dt$$

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$f$  of bandwidth  $\leq \Omega \Rightarrow$

$$\hat{f}(\omega) = \Delta t \sum_{n=-\infty}^{\infty} f_n e^{-2\pi i n \Delta t \omega} \quad \forall \omega \in [-\Omega, +\Omega].$$

$\Delta t = \frac{1}{2\Omega}$  is the **Nyquist rate**.

## Theorem.

$f$  of bandwidth  $\leq \Omega$  & sample frequency  $1/\Delta t \geq 2\Omega \Rightarrow$

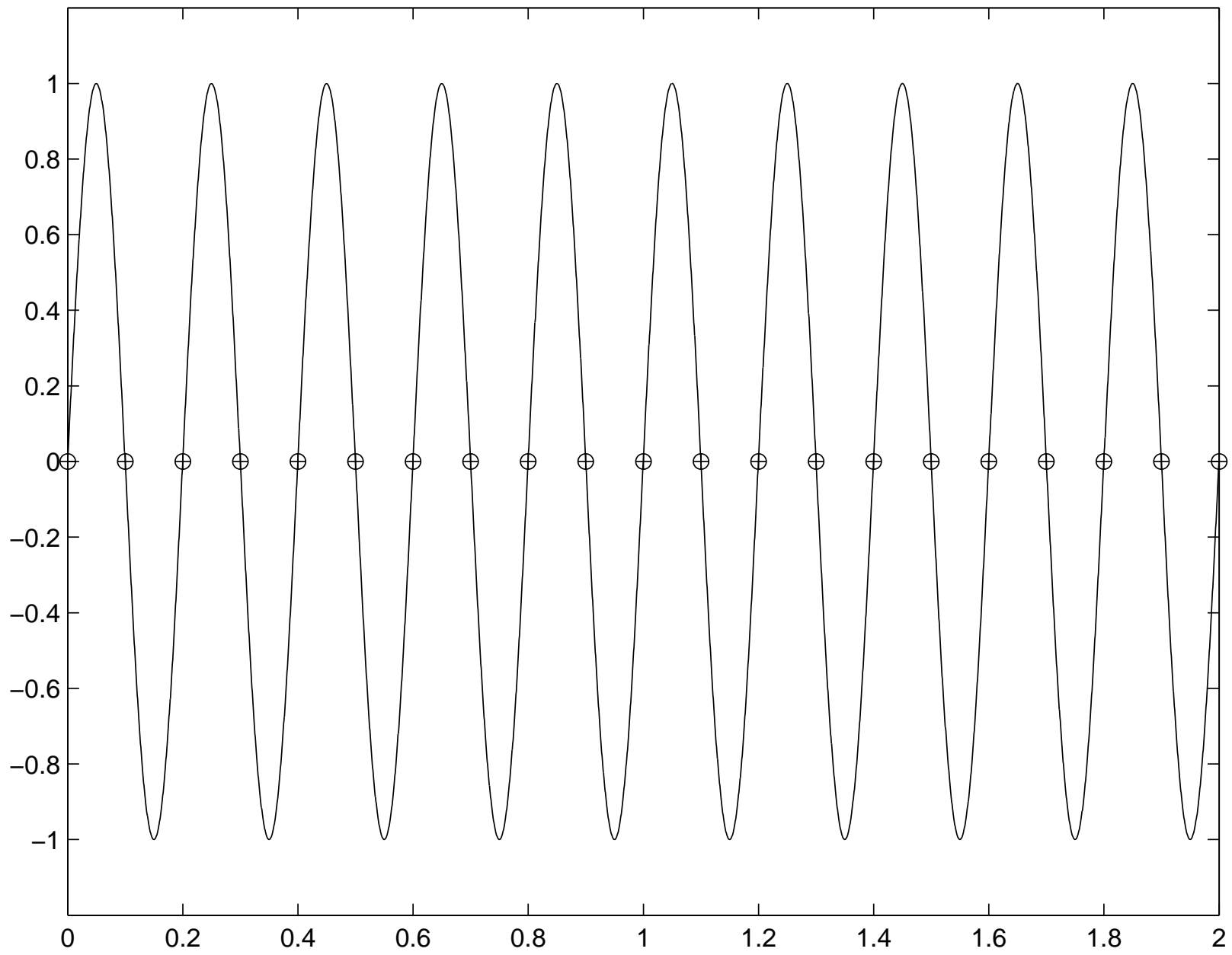
$$\hat{f}(\omega) = \Delta t \sum_{n=-\infty}^{\infty} f_n e^{-2\pi i n \Delta t \omega} \quad \forall \omega \in [-\Omega, +\Omega]$$

The discretization is exact if the bandwidth  $\leq \Omega$  and the sample frequency  $\geq 2\Omega$  ( $\Delta t \leq 1/(2\Omega)$ ). Fourier transform of this result leads to

## The Shannon–Whittaker Theorem.

$f$  of bandwidth  $\leq \Omega$  & sample frequency  $1/\Delta t \geq 2\Omega \Rightarrow$

$$f(t) = \sum_{n=-\infty}^{\infty} f_n \operatorname{sinc}\left(\frac{t - t_n}{\Delta t}\right) \quad \forall t \in \mathbb{R}.$$



**Discussion.** The Shannon–Whittaker theorem tells us that  $f$  can be reconstructed from its sample values, if  $f$  is of bounded bandwidth and the sample frequency is at least twice the maximal frequency of  $f$ . However, reconstruction requires values  $f_n$  from the (far) future as well as from the (far) past.

**Application.** Resampling (sampling at another sampling rate) is possible.

If the new sample rate is  $\frac{p}{q}$  times the old sample rate  $\Delta t$ , then, in practice, resampling is achieved by

- 1) upsampling by  $p$
- 2) filtering to get rid of frequencies  $> \Omega$
- 3) downsampling by  $q$ .

(Details later)

## Conclusions

- Discretization is fine provided  $f$  is of bounded bandwidth and the sample frequency is high enough.
- Perturbations by windowing can not be avoided. Effects include smearing and leakage. Effects can be diminished by a larger time-window. One effect can be diminished at the cost of others (by other time-windows).